

# Localization Cues Preservation in Hearing Aids by Combining Noise Reduction and Dynamic Range Compression

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## Introduction

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# Dynamic range compression (DRC)

**Aim:** amplify the soft sound without reaching the pain threshold

Dynamic Range Compression (DRC):

- Attenuate the output if the level exceeds a given threshold
- Number of frequency bands : 8  $\rightarrow$  32

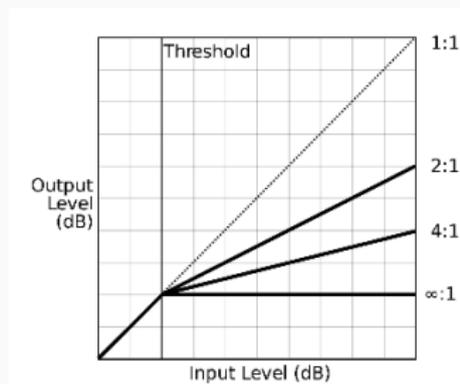


Figure 1: DRC broken-stick function

Issue: left and right DRC gain are different:

- Hearing loss compensation distorts the localization cues
- Localization performance decreasing<sup>1</sup>
- Speech in noise understanding performance decreasing<sup>2</sup>

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<sup>1</sup>[Hassager et al., 2017b, Wiggins and Seeber, 2011, Van den Bogaert et al., 2006]

<sup>2</sup>[Schwartz and Shinn-Cunningham, 2013]

- **Aims :**
  - reducing the speech dynamic range
  - preserving the original noise dynamic range
  - improving the output SNR
- **Idea:** fast compression for the speech period, slowly otherwise

## Advantage

improve the localization in presence of reverberation [Hassager et al., 2017a]

## Drawbacks

The attenuation of the noise only period depends on the previous speech content!

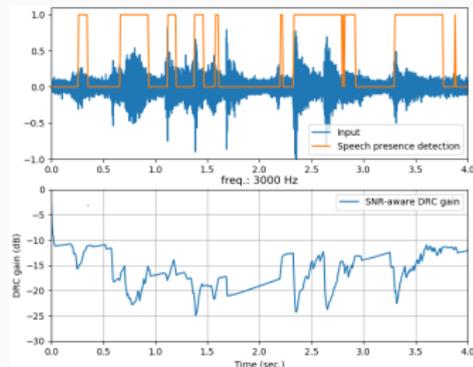


Figure 2: Noisy speech and speech presence detection (top), DRC gain at 3 kHz of the SNR-aware DRC

<sup>3</sup>[May et al., 2018]

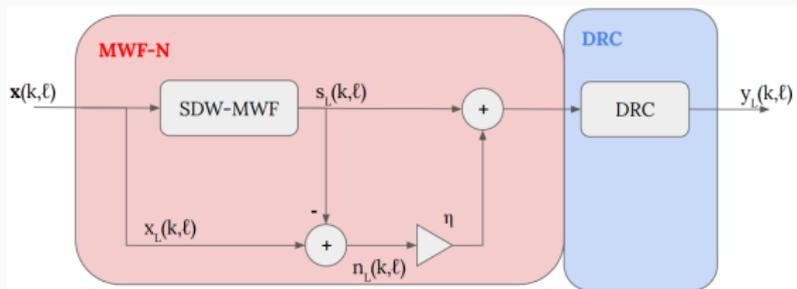


Figure 3: Standard association of noise reduction and DRC.

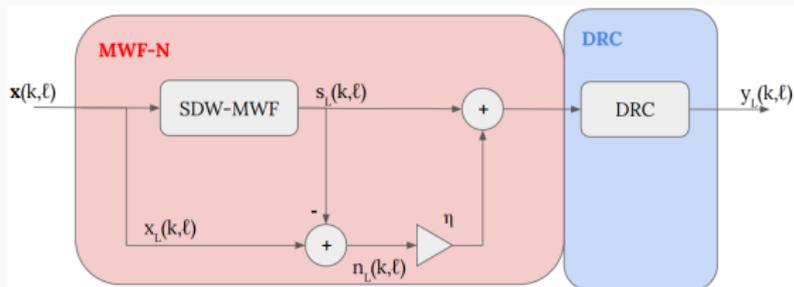


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## Objectives :

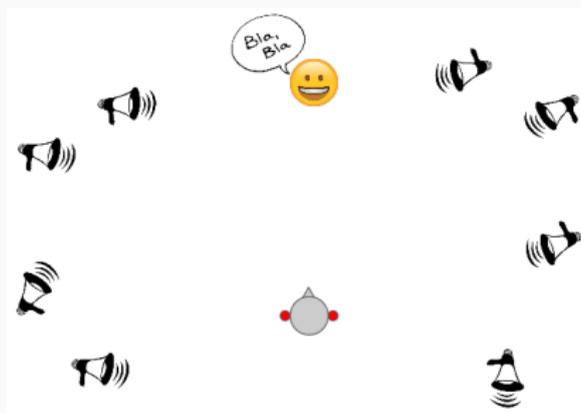
- Reducing the speech dynamic range
- Preserving the noise dynamic range
- Improving the SNR
- Preserving the localization cues of both components

## Idea

Merge noise reduction and DRC

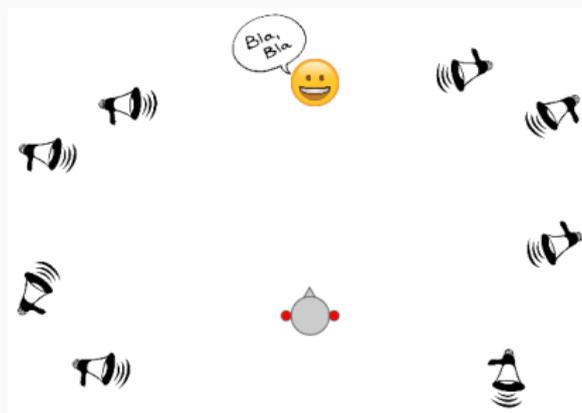
## Data Model

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At each microphone  $x_m(t)$ :

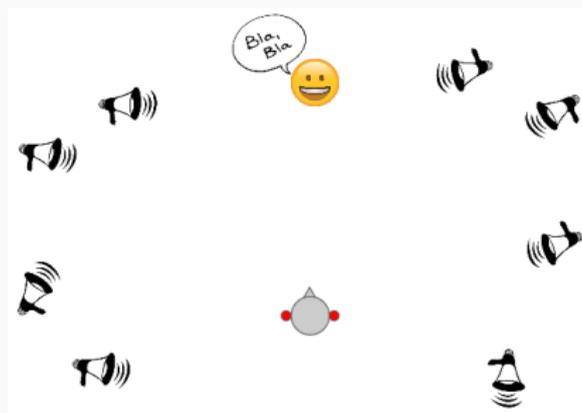
- one speech source (plane wave),  $s(t)$ , filtered by  $h_m(t)$ .
- a noise component (spatially diffuse),  $n_m(t)$ .



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$$\text{Time domain: } x_m(t) = (h_m \star s)(t) + n_m(t) \quad (1)$$

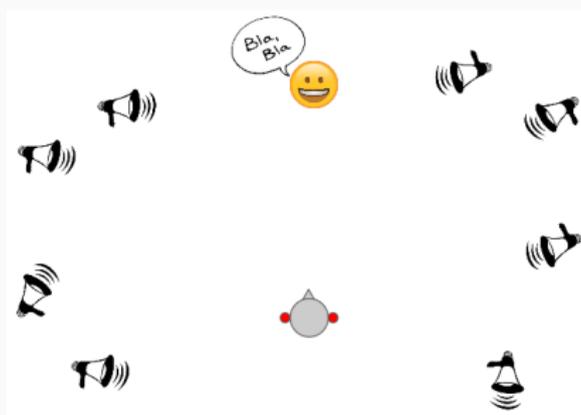


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$$\text{Time domain: } x_m(t) = (h_m \star s)(t) + n_m(t) \quad (1)$$

$$\text{STFT domain: } x_m(k, \ell) = h_m(k, \ell)s(k, \ell) + n_m(k, \ell) \quad (2)$$



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$$\text{STFT domain: } x_m(k, \ell) = h_m(k, \ell)s(k, \ell) + n_m(k, \ell) \quad (2)$$

$$\text{matrix notation: } \mathbf{x}(k, \ell) = \mathbf{h}(k) s(k, \ell) + \mathbf{n}(k, \ell). \quad (3)$$

$\mathbf{x}(k, \ell) \in \mathbb{C}^M$ ,  $\mathbf{h}(k, \ell) \in \mathbb{C}^M$  and  $\mathbf{n}(k, \ell) \in \mathbb{C}^M$ .

Speech sparsity assumption in the STFT domain:

- $\mathcal{H}_0: \mathbf{x}(k, \ell) = \mathbf{n}(k, \ell),$
- $\mathcal{H}_1: \mathbf{x}(k, \ell) = \mathbf{h}(k)\mathbf{s}(k, \ell) + \mathbf{n}(k, \ell)$

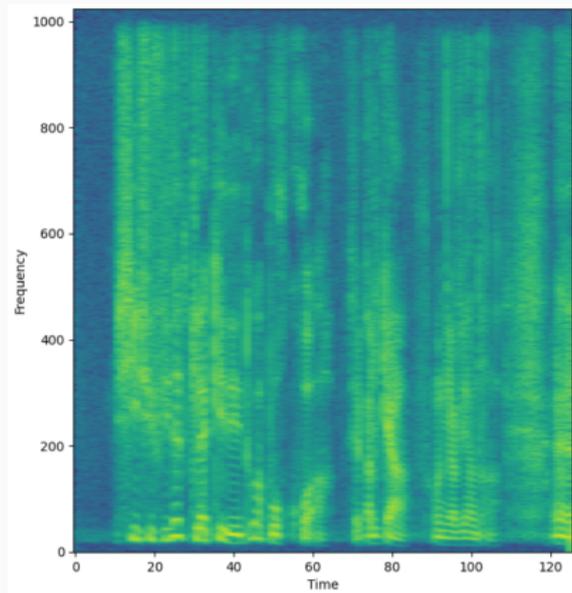


Figure 4: Speech spectrogram

Discrete Fourier coefficients are modeled as random variables:

$$s(k, \ell) \sim \mathcal{N}_{\mathbb{C}}(0, \phi_s(k, \ell)) \quad (4)$$

$$\mathbf{n}(k, \ell) \sim \mathcal{N}_{\mathbb{C}}(0, \Phi_{nn}(k, \ell)) \quad (5)$$

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Noise covariance matrix model (spatially diffuse):

$$\Phi_{nn}(k, \ell) = \phi_n(k, \ell) \Gamma_{\text{diff}}(k) \quad (6)$$

## Proposed Algorithm

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$\hat{s}_L(k, \ell)$ : speech source estimate at the left ear (same for right ear)

$$\hat{s}_L(k, \ell) = \mathbf{w}^H(k, \ell) \mathbf{x}(k, \ell) \quad (7)$$

with  $\mathbf{w}(k, \ell) \in \mathbb{C}^M$ .

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Optimization problem:

$$\hat{\mathbf{w}}_L(k, \ell) = \underset{\mathbf{w}}{\operatorname{argmin}} \left\{ \mathbb{E} \left[ \left| s_L(k, \ell) - \mathbf{w}^H \mathbf{x}(k, \ell) \right|^2 \right] \right\} \quad (8)$$

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Using the speech sparsity assumption

$$\begin{aligned} \mathbb{E} \left[ \left| s_L(k, \ell) - \mathbf{w}^H \mathbf{x}(k, \ell) \right|^2 \right] &= P(\ell) \left[ p(k, \ell) \mathbb{E} \left[ \left| s_L(k, \ell) - \mathbf{w}^H \mathbf{x}(k, \ell) \right|^2 \mid \mathcal{H}_1 \right] \right. \\ &\quad \left. + (1 - p(k, \ell)) \mathbb{E} \left[ \left| \mathbf{w}^H \mathbf{x}(k, \ell) \right|^2 \mid \mathcal{H}_0 \right] \right] \\ &\quad + (1 - P(\ell)) \left[ \frac{1}{\mu_{\mathcal{H}_0}} \mathbb{E} \left[ \left| s_L(k, \ell) - \mathbf{w}^H \mathbf{h}(k) s(k, \ell) \right|^2 \right] \right. \\ &\quad \left. + \mathbb{E} \left[ \left| \mathbf{w}^H \mathbf{x}(k, \ell) \right|^2 \right] \right] \end{aligned} \quad (9)$$

Informed Multichannel Wiener Filter (MWF) for the left ear:

$$\hat{\mathbf{w}}_L(k, \ell) = \left( \phi_s(k, \ell) \mathbf{h}(k) \mathbf{h}(k)^H + \mu(k, \ell) \Phi_{nn}(k, \ell) \right)^{-1} \mathbf{h}(k) \phi_s(k, \ell) h_L(k)^*, \quad (10)$$

with

$$\mu(k, \ell) = P(\ell) \frac{1}{p(k, \ell)} + (1 - P(\ell)) \mu_{\mathcal{H}_0}. \quad (11)$$

- $p(k, \ell)$ : narrowband speech presence probability estimation
- $P(\ell)$ : broadband speech presence probability estimation

Idea introduced by [Ngo et al., 2012] and similarly by [May et al., 2018]: use different DRC to process speech source and noise source

- $DRC_s$ : to process the speech source when it is active at this T-F bin,
- $DRC_n H_1$ : to process the speech source when it is NOT active at this T-F bin,
- $DRC_n H_0$ : to process the speech source when it is NOT active at all bins,
- $DRC_n$ : to process the noise component.

DRC	Attack (ms)	Release (ms)	Gain $G_0$ (dB)
$DRC_s$	10	60	0
$DRC_n H_1$	10	2000	-6
$DRC_n H_0$	10	2000	-10
$DRC_n$	2000	2000	-10

Multichannel estimator [Souden et al., 2010]

$P(\mathcal{H}_1|\mathbf{x}(k, \ell))$ : *a posteriori* source presence probability, denoted  $p(k, \ell)$

Bayes rule:

$$P(\mathcal{H}_1|\mathbf{x}(k, \ell)) = \frac{P(\mathbf{x}(k, \ell)|\mathcal{H}_1)P(\mathcal{H}_1)}{P(\mathbf{x}(k, \ell)|\mathcal{H}_1)P(\mathcal{H}_1) + P(\mathbf{x}(k, \ell)|\mathcal{H}_0)P(\mathcal{H}_0)} \quad (12)$$

$P(\mathbf{x}(k, \ell)|\mathcal{H}_1)$ : data likelihood according to the Gaussian assumption

$P(\mathcal{H}_1)$ : prior (adaptive [Cohen, 2002])

Broadband binary detector

$$\hat{P}(\ell) = \begin{cases} 1 & \text{if } \sum_k p(k, \ell) > t_{high} \text{ and } P(\ell - 1) = 0 \\ 0 & \text{if } \sum_k p(k, \ell) > t_{low} \text{ and } P(\ell - 1) = 1 \\ P(\ell - 1) & \text{otherwise,} \end{cases} \quad (13)$$

Recursive filtering

$$P(\ell) = \alpha_P \hat{P}(\ell) + (1 - \alpha_P) P(\ell - 1) \quad (14)$$

- Similar to the proposition of [Ngo et al., 2012]
- Improvements :
  - DRC association more consistent with the literature
  - attack and release time constant decorrelation between DRC and broadband speech detection
  - binaural rather than monaural

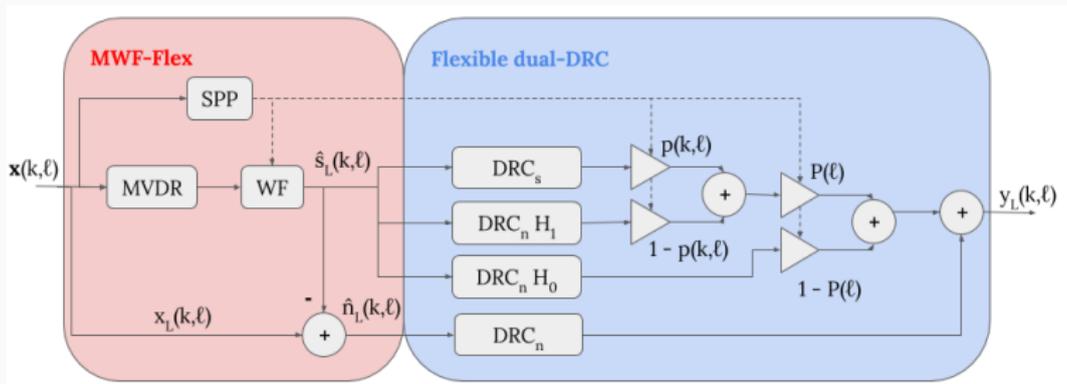
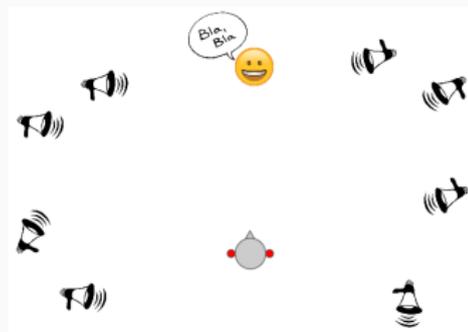


Figure 5: Our proposition

## Experiments

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- Scenario:
  - speaker located in front of the listener
  - cafeteria noise
  - SNR: 5 dB
- Ideal scenario: SNR of 15 dB
- With 14 different HRTFs



Interaural coherence

$$IC = \max_{\tau} \left| \frac{\sum_t \tilde{y}_L(t + \tau) \tilde{y}_R(t)}{\sqrt{\sum_t |\tilde{y}_L(t)|^2 \sum_t |\tilde{y}_R(t)|^2}} \right|. \quad (15)$$

## Advantages

- Denoising across all the pipeline
- Interaural coherence closer to the ideal scenario

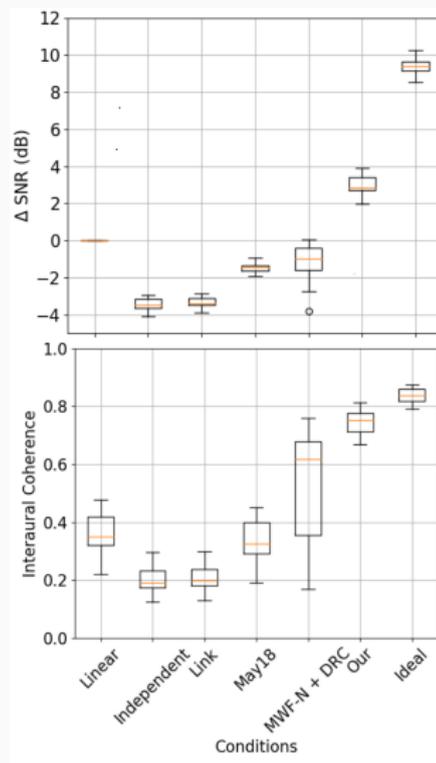


Figure 6: SNR improvement (top) and interaural coherence (bottom).

Effective Compression Ratio (ECR):

- $< 1$ : more dynamic range
- $> 1$ : less dynamic range

### Drawbacks

- Speech cut  $\rightarrow$  ECR deteriorated
- Residual noise component into the speech branch correlated with the speech  $\rightarrow$   $ECR < 1$

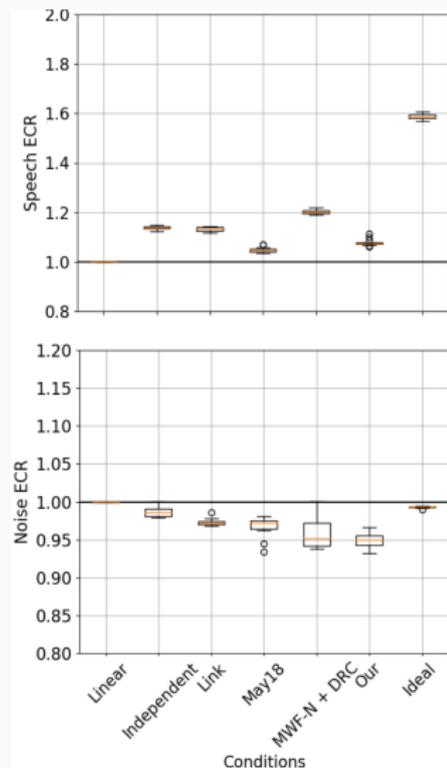


Figure 7: Speech ECR (top) and noise ECR (bottom).

- SNR improvement
- Better interaural coherence
- ECR deterioration due to estimation errors
- Future works: perceptive test

Please share your comments and questions

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